

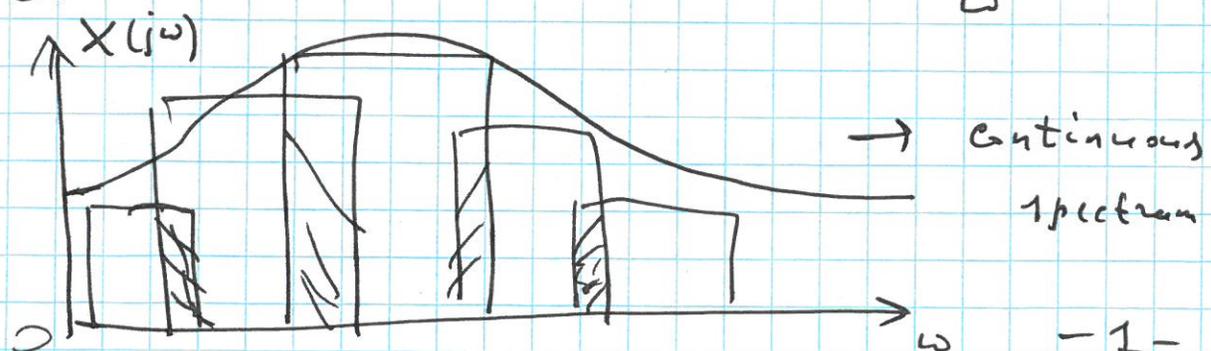
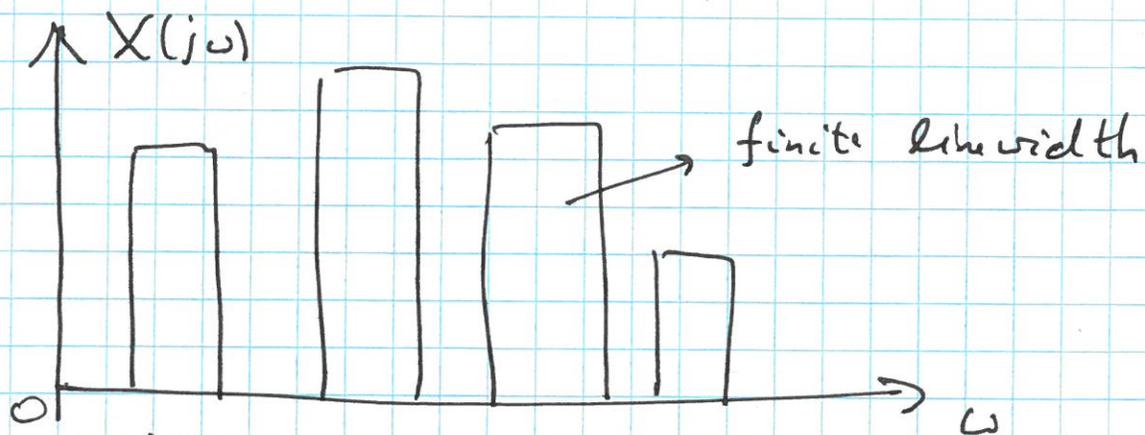
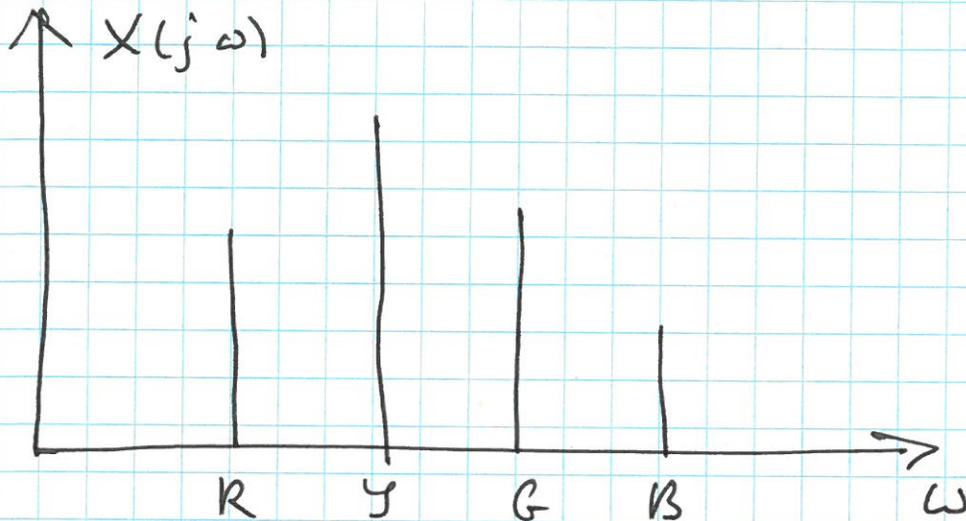
1. Fourier Transforms

Periodic function can always be expanded into a finite number of harmonics $e^{+jn\omega}$

Example:

$$\cos \omega t = \frac{1}{2}(e^{j\omega t} + e^{-j\omega t})$$

Nature example of an FT: rainbow



$$\sum_n X_{nw} e^{j\omega t} \rightarrow \int dX(\omega) e^{j\omega t} \rightarrow$$

$$\rightarrow \int \frac{d\omega}{2\pi} X(j\omega) e^{j\omega t}$$

↓
spectral density of signal $X(t)$

∴

$$X(t) = \int \frac{d\omega}{2\pi} X(j\omega) e^{j\omega t} \quad (\text{inverse FT})$$

$$X(j\omega) = \int dt X(t) e^{-j\omega t} \quad (\text{direct FT})$$

Properties:

1) Shift $X(t-t_0) \xleftrightarrow{\text{FT}} e^{-j\omega t_0} X(j\omega)$

Consider $X(j\omega) = \int dt X(t) e^{-j\omega t}$

Take $\int dt X(t-t_0) e^{-j\omega t} \quad | \quad \tau = t-t_0 \quad | =$

$$= \underbrace{\int d\tau X(\tau) e^{-j\omega \tau}}_{X(j\omega)} e^{-j\omega t_0} = e^{-j\omega t_0} X(j\omega)$$

2) Complex conjugation

$$\mathcal{F}[x^*(t)] = X^*(-j\omega)$$

$$x^*(t) = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X^*(j\omega) e^{-j\omega t} \quad \left| \omega \rightarrow -\omega \right| =$$

$$= \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X^*(-j\omega) e^{j\omega t}$$

$$\therefore \mathcal{F}[x^*(t)] = X^*(-j\omega)$$

3) Derivative

$$x'(t) \equiv \frac{dx}{dt} = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X(j\omega) \frac{d}{dt} e^{j\omega t} = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} j\omega X(j\omega) x e^{j\omega t}$$

$$\therefore \mathcal{F}[x'(t)] = j\omega X(j\omega)$$

4) Parseval theorem.

Introduce signal energy

$$E = \int_{-\infty}^{\infty} dt |x(t)|^2$$

$$E = \int_{-\infty}^{\infty} dt |x(t)|^2 = \int_{-\infty}^{\infty} dt x^*(t) x(t) = \int_{-\infty}^{\infty} dt x^*(t) x$$

$$x \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X(j\omega) e^{j\omega t}$$

Convolutions describe response of linear, time-invariant systems (LTI) to external signal inputs

Examples: 1. Capacitor

$$q = Cv, \quad q = \int_0^t dt' i(t')$$

$$v = \frac{1}{C} q = \frac{1}{C} \int_0^t dt' i(t')$$

$$v = \int_{-\infty}^{\infty} d\tau i(\tau) h(t-\tau)$$

$$\underbrace{h(t) = \frac{1}{C} u(t)}, \quad u(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

2. Inductor

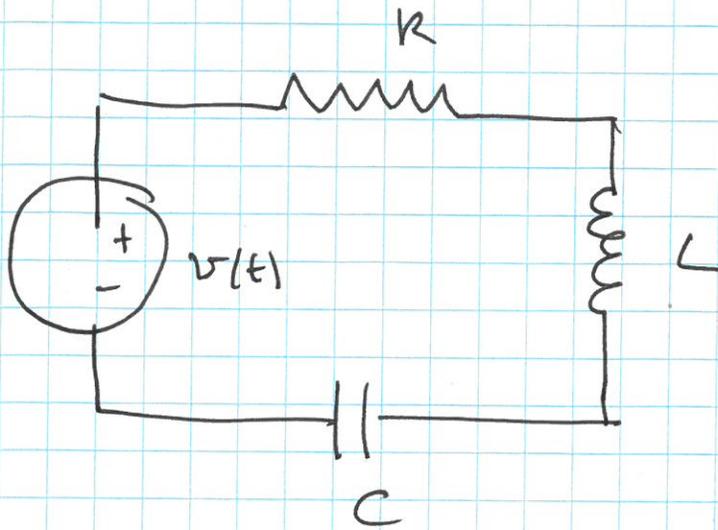
$$v = L \frac{di}{dt}$$

$$v = L \int_{-\infty}^{\infty} d\tau i(\tau) \left[-\frac{d}{dt} \delta(t-\tau) \right]$$

\therefore

$$\underbrace{h(t) = -L \frac{d}{dt} \delta(t)}$$

RLC - circuit response with FT



$$\frac{q}{c} + R \dot{q} + L \ddot{q} = v(t), \quad \text{Kirchhoff}$$

$$\text{FT:} \quad Q(j\omega) = \int_{-\infty}^{\infty} dt q(t) e^{-j\omega t}$$

$$V(j\omega) = \int_{-\infty}^{\infty} dt v(t) e^{-j\omega t}$$

$$\frac{Q}{c} + j\omega R Q - \omega^2 Q L = V$$

$$(-\omega^2 + 1/Lc + j \frac{R}{L} \omega) Q = V/L$$

Introduce $2\beta = R/L, \quad \omega_s^2 = 1/Lc$

$$\therefore (-\omega^2 + \omega_s^2 + 2j\beta\omega) Q = V/L$$

$$Q(j\omega) = H(j\omega) V(j\omega)$$

\therefore

$$H(j\omega) = \frac{1}{L(\omega_s^2 - \omega^2 + 2j\gamma\omega)}$$

Consider a good quality LRC: $\gamma \ll \omega_s$

Take ω to be close to ω_s : $\omega \approx \omega_s$

$$\therefore \omega \approx \omega_s, \quad \gamma \ll \omega_s$$

\therefore

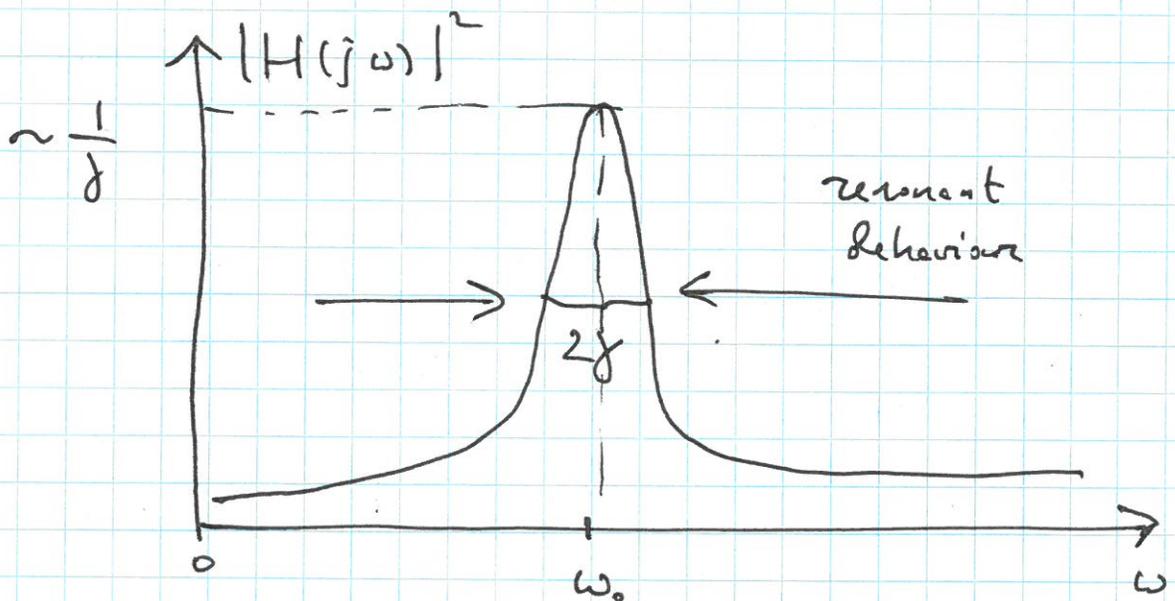
$$\begin{aligned} \omega_s^2 - \omega^2 + 2j\gamma\omega &= (\omega_s - \omega)(\omega_s + \omega) + 2j\gamma\omega \approx \\ &\approx 2\omega_s(\omega_s - \omega) + 2j\gamma\omega_s \approx 2\omega_s(\omega_s - \omega + j\gamma) \end{aligned}$$

\therefore

$$H(j\omega) \approx \frac{1}{2L\omega_s(\omega_s - \omega + j\gamma)}$$

\therefore

$$|H(j\omega)|^2 \approx \frac{1}{4L^2\omega_s^2 [(\omega_s - \omega)^2 + \gamma^2]}$$



Laplace transform

Definiere:

$$X(s) = \int_{-\infty}^{\infty} dt x(t) e^{-st}$$
$$s = \sigma + j\omega$$

Inverse:

$$X(s) = \int_{-\infty}^{\infty} dt x(t) e^{-\sigma t} e^{-j\omega t}$$
$$\tilde{x}(t)$$

$$\tilde{x}(t) \equiv x(t) e^{-\sigma t} = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X(\sigma + j\omega) e^{j\omega t}$$

$$\therefore x(t) = \int_{-\infty}^{\infty} \frac{d\omega}{2\pi} X(\sigma + j\omega) e^{(\sigma + j\omega)t}$$

$$\therefore x(t) = \int_{\sigma - j\infty}^{\sigma + j\infty} \frac{ds}{2\pi j} X(s) e^{st}$$

or

$$x(t) = \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} ds X(s) e^{st}$$

Likely properties:

1) shift: $\mathcal{L}[x(t-t_0)] = e^{-st_0} X(s)$

2) conjugation $\mathcal{L}[x^*(t)] = X^*(s^*)$

3) convolution $\mathcal{L}[x_1 * x_2] = X_1(s) X_2(s)$

4) derivative $\mathcal{L}\left[\frac{d}{dt} x\right] = s X(s)$

5) Finite value theorem

Suppose the signal $x(t)$ is switched on at $t=0$:

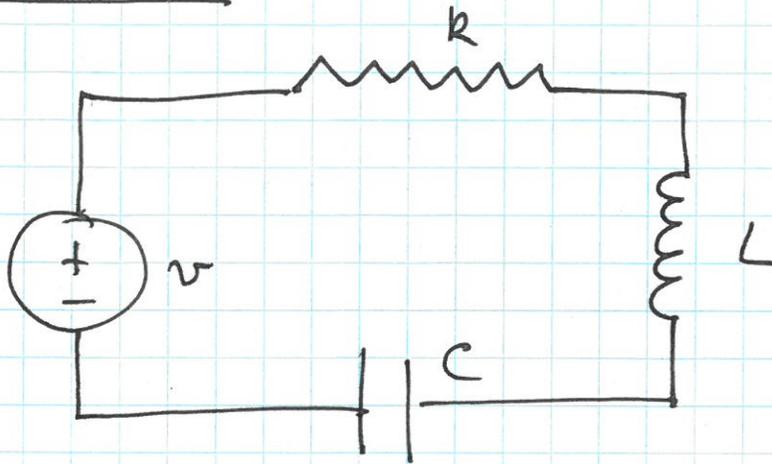
$$x(t) = \begin{cases} x(t), & t > 0 \\ 0, & t \leq 0 \end{cases}$$

and there exists a finite limit

$$X_\infty = \lim_{t \rightarrow \infty} x(t) < \infty,$$

$$X_\infty = \lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} s X(s)$$

Example 1 RLC-circuit twisted



$$L \ddot{q} + R \dot{q} + \frac{1}{C} q = v$$

$$Q(s) = \int_{-\infty}^{\infty} dt q(t) e^{-st}$$

$$V(s) = \int_{-\infty}^{\infty} dt v(t) e^{-st}$$

$$(s^2 L + R s + 1/C) Q = V$$

∴

$$Q = \frac{V/L}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$\omega_0^2 = 1/LC, \quad 2\gamma = R/L$$

∴

$$Q = \frac{V/L}{s^2 + 2\gamma s + \omega_0^2}$$

$$Q = \frac{V/L}{(s + \gamma)^2 + \omega_0^2 - \gamma^2}$$

$$Q(s) = H(s) V(s)$$

\therefore

$$H(s) = \frac{1/L}{(s+\gamma)^2 + \omega_0^2 - \gamma^2}$$

Use tables:

$$\mathcal{L}^{-1} \left[\frac{v}{(s+d)^2 + v^2} \right] = u(t) e^{-dt} \sin vt$$

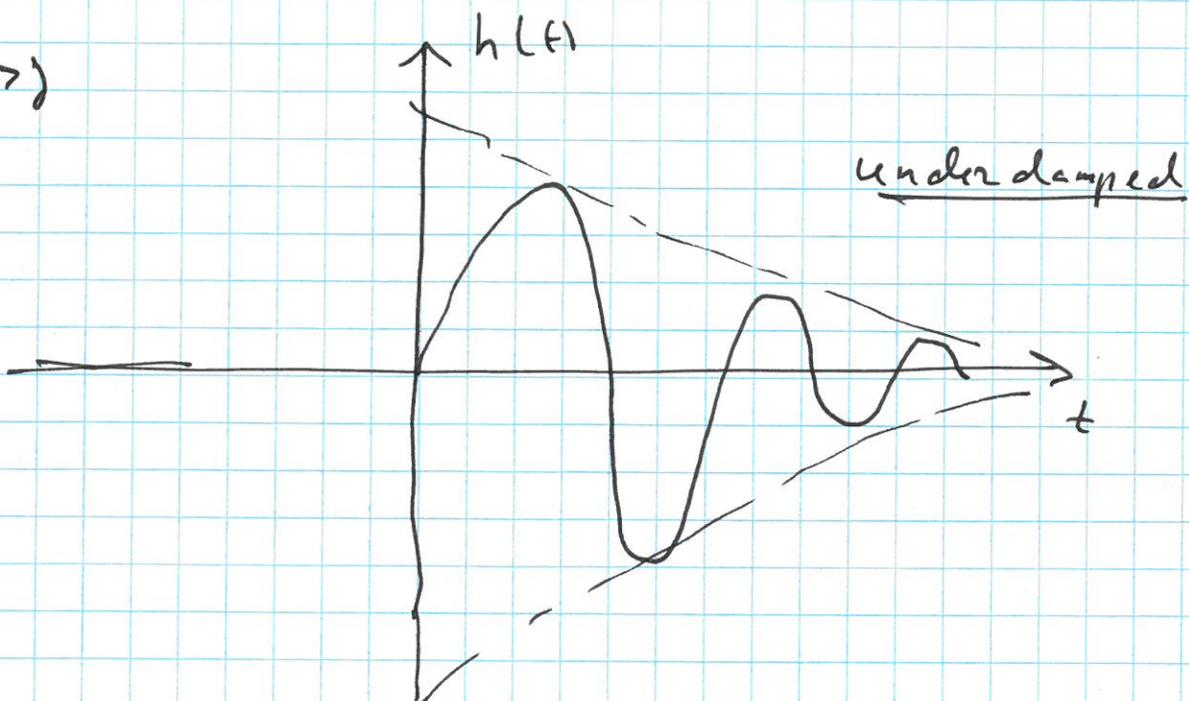
\therefore

$$h(t) = \mathcal{L}^{-1} u(t) e^{-\gamma t} \frac{\sin(\sqrt{\omega_0^2 - \gamma^2} t)}{\sqrt{\omega_0^2 - \gamma^2}}$$

\therefore

$$h(t) = u(t) e^{-\gamma t} \frac{\sin(t \sqrt{\omega_0^2 - \gamma^2})}{\sqrt{\omega_0^2 - \gamma^2}}$$

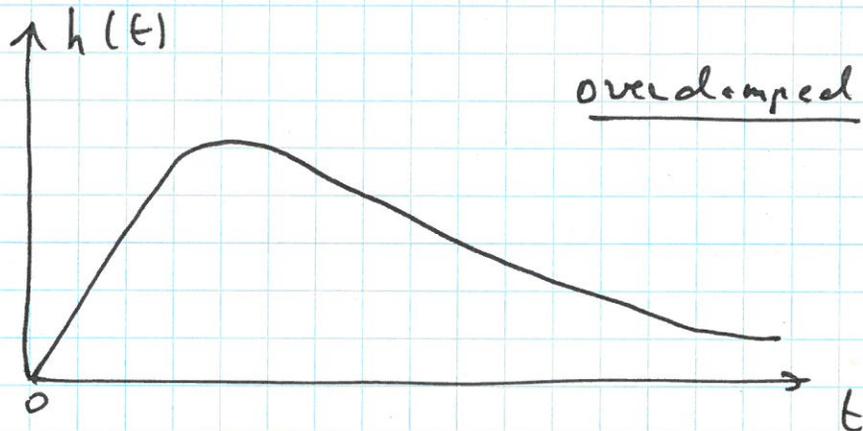
a) $\omega_0 > \gamma$



b) $\omega_0 < \delta \quad \sin jx \rightarrow j \sinh x$

$$\sqrt{\omega_0^2 - \delta^2} = j \sqrt{\delta^2 - \omega_0^2}$$

$$h(t) \rightarrow u(t) e^{-\delta t} \frac{\sinh(t \sqrt{\delta^2 - \omega_0^2})}{\sqrt{\delta^2 - \omega_0^2}}$$

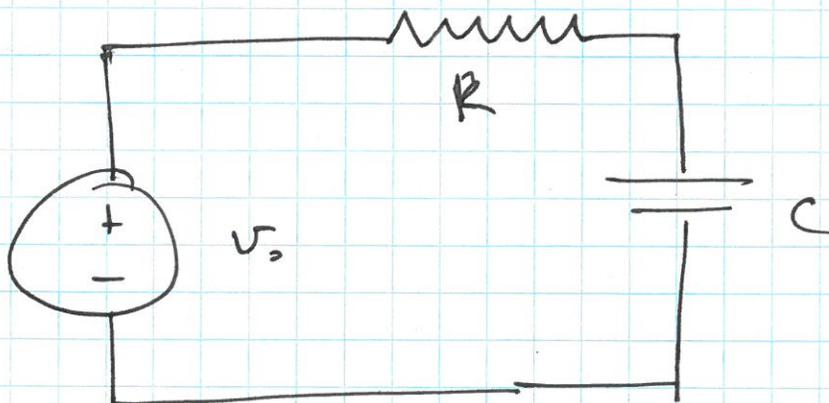


c) $\omega_0 = \delta \quad h(t) = u(t) e^{-\delta t} \lim_{\omega_0 \rightarrow \delta} \frac{\sin(t \sqrt{\omega_0^2 - \delta^2})}{\sqrt{\omega_0^2 - \delta^2}}$

$$\approx u(t) t e^{-\delta t}$$

critically damped response

Example 2 RC-circuit asymptotics



$$R \dot{q} + \frac{1}{C} q = V_s$$

$$\dot{q} + \frac{1}{\tau} q = V_s/R, \quad \tau = RC$$

$$sQ + \frac{1}{\tau} Q = \frac{V_s}{sR}$$

-:

$$Q(s) = \frac{V_s/R}{s(s + 1/\tau)}$$

$$Q_\infty = \lim_{s \rightarrow 0} sQ(s) = \frac{V_s/R}{1/\tau} = CV_s$$

The capacitor will charge to the source ~~voltage~~
voltage!